Distributed coordinated active and reactive power control of wind farms based on model predictive control

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Distributed Coordinated Active and Reactive Power Control of Wind Farms
Based on Model Predictive Control

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Abstract

This paper proposes a distributed coordinated active and reactive power control scheme for wind farms based on the model predictive control (MPC) along with the consensus-based distributed information synchronization and estimation, which can optimally dispatch the active power of wind turbines (WTs) and regulate the voltages within the wind farm. For the active power control, the pitch angle and generator torque of WTs are optimally controlled to alleviate fatigue loads of WTs while tracking the power reference of the wind farm required by system operators. For the reactive power/voltage control, the reactive power outputs of WTs are controlled to mitigate the voltage deviations and simultaneously optimize reactive power sharing. Considering the high $R/X$ ratio of the wind farm collector systems, the impact of active power variations on voltages is taken into account to improve the voltage regulation. The proposed scheme is center-free and only requires sparse communication network. Each WT only exchanges information with its immediate neighbors and the local optimal control problems are solved in parallel, implying good scalability and flexibility for large-scale wind farms. The predictive model of a WT is derived and then the MPC problem is formulated. A wind farm with ten WTs was used to verify the proposed control scheme.

Keywords: active power control, consensus protocol, distributed control, model predictive control, reactive power control, voltage control, wind farm.

Nomenclature

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<th>Symbol</th>
<th>Description</th>
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<td>$(*)_0$</td>
<td>measurement at operating point</td>
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<td>$\Delta(*)$</td>
<td>incremental value of variable</td>
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<td>$(*)^{(i)}$</td>
<td>variables estimated by WT-i</td>
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<tr>
<td>$P_{WT}^{ref}$</td>
<td>active power reference of WT</td>
</tr>
<tr>
<td>$P_g$</td>
<td>active power output of WT</td>
</tr>
<tr>
<td>$\omega_r$</td>
<td>rotor speed</td>
</tr>
<tr>
<td>$\omega_{r}, \omega_f$</td>
<td>generator and filtered speed</td>
</tr>
<tr>
<td>$\theta, \theta_{ref}$</td>
<td>pitch angle and its reference</td>
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<tr>
<td>$T_e$</td>
<td>aerodynamics torque</td>
</tr>
<tr>
<td>$F_t$</td>
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<td>reactive power output of WT</td>
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<td>reactive power reference of WT</td>
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<td>available power of wind farm</td>
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<td>$P_{WT}^{av}$</td>
<td>average available power of WTs</td>
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<tr>
<td>$V_{ref}$</td>
<td>voltage reference</td>
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<tr>
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<td>voltage at the POC</td>
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1. Introduction

Wind energy has been developing rapidly due to the growing concerns over environmental issues around the world. As the wind power penetration level increases, its variability and uncertainty have brought a number of technical and economic challenges for power system operation [1]. Wind farms are required to meet the technical requirements specified in grid codes issued by system operators. With the fast development of modern wind turbine (WT) technologies, the controllability and fast-response capability of wind power are significantly improved. Modern wind farms are able to provide multiple ancillary services such as grid frequency and voltage support [2], [3].

Generally, wind farm control applications may consist of several control objectives including active power dispatch and reactive power/voltage control [4]. Due to the decoupled control loops of active and reactive power of modern WTs, they are often separately designed.

For active power control, several control strategies such as proportional distribution [5]-[6] and proportional-integral (PI) control [7] are easy to be implemented. However, these dispatch methods mainly focus on power reference tracking of the wind farm without considering fatigue loads experienced by the WTs, which have significant impacts on the lifetime of WTs. In recent years, the optimization-based control strategies have been widely studied [8]-[10]. The optimal control problems are formulated as multi-objective optimization problems which can achieve power tracking as well as reduce the fatigue loads of WTs.

For reactive power control of wind farms, the main aim is to maintain the voltage at the point of connection (POC) within the feasible range, which is specified in many grid codes. Several reactive power dispatch strategies based on the proportional distribution and PI controller have been proposed in [4], [10]-[11], which depend on the voltage at the POC and available reactive power capability of WTs while the terminal voltages of WTs are not considered. The optimization-based reactive power/voltage control strategies have also been proposed [12]-[14]. In [12], a hierarchical automatic voltage controller based on the sensitivity method was designed and implemented in a wind power base of northern China. In [13]-[14], the optimal power flow-based control strategies for high-voltage-direct-current (HVDC) connected offshore wind farms were proposed, in which the objectives are to minimize the active power losses of the offshore system.

As a special optimization-based method, model predictive control (MPC) has been widely used in wind power systems both in WT level [15]-[16] and wind farm level [17]-[22]. In [17], a MPC scheme was proposed to balance the wind farm power reference tracking as well as fatigue loads reduction. In [18]-[20], the distributed MPC (D-MPC) schemes were proposed for optimal active power dispatch for wind farms, in which the optimal control problems are solved by the distributed optimization algorithms, however, in which a central unit is also required for WT coordination to track the power reference required by system operators. In [21]-[22], the centralized MPC-based coordinated wind farm voltage control schemes were designed. In [21], the reactive power sources inside a wind farm including WTs, static Var compensators and on-load tap changing transformer are optimally coordinated. In [22], WTs and wind farm side HVDC converter are optimally coordinated.

In the centralized optimization-based voltage control schemes [12]-[14], [17]-[22], the wind farm is modeled as a constrained multiple input and multiple output system whose order drastically grows as the number of WTs increases. As the number of WTs increases, the computation burden of the central controller will be heavy. Moreover, the cost of the communication infrastructures may be quite high for large-scale wind farms. Distributed control is appealing for the wind farm control since a wind farm consists of a number of WTs, which has a high degree autonomy. Besides, the typical \( R/X \) ratio of the wind farm collector system is high and consequently voltages are sensitive to variations in active power injections. The conventional active and reactive power control schemes of wind farms were designed in a separated manner, which neglects the impact of active power variations of WTs on system voltages and consequently voltage control performance might not be optimal.

In this context, the main contribution of this paper is a distributed coordinated active and reactive power control design which aims to optimally regulate active and reactive power outputs of WTs in a wind farm. For the active power control, the controllers reduce the fatigue loads of WTs while tracking the wind farm power reference. For the reactive power control, the voltages are regulated and reactive power sharing is optimized. The impact of active power variations on voltages is taken into account to improve the voltage control. The global reference information including the power reference of the wind farm and voltage at the POC is synchronized by a distributed finite-time observer. The total available active power of the wind farm is estimated using a distributed estimator based on the average-consensus protocol.
Compared with the traditional centralized control, the proposed distributed control scheme has several advantages as follows: 1) Eliminate the requirement of a central controller; 2) Reduce the cost of communication infrastructures (All WTs only exchange information with their immediate neighbors, and when the wind farm is expanded, the newly connected WTs are only required to build the communication link with the neighboring WTs instead of the far central controller); 3) Reduce the computation burden. (the computation in each WT controller is in parallel and the computation burden is less dependent on the number of WTs); 4) Improve the scalability of the wind farm controller.

The rest of this paper is organized as follows. Section 2 presents an overview of the proposed control scheme. In Section 3, the predictive models of WTs regarding active and reactive power control loops are described. In Section 4, the formulation of the distributed optimal control problem is presented. Simulation results are presented and discussed in Section 5 followed by conclusions.

2. Distributed MPC Based Power Control Scheme Design for Wind Farms

The typical structure of a wind farm is shown in Fig. 1. Each WT is equipped with a distributed WT controller and a sparse, connected communication network is designed for the distributed power control scheme.

The configuration of the proposed distributed coordinated power controller (DCPC) is illustrated in Fig. 2. The exchanged information between the system operator (or wind farm operator) and the leader WT is the wind farm power reference. The information between two neighboring WTs includes the estimated available power of the wind farm, the estimated power reference of wind farm, the estimated voltage at the POC and the predicted active/reactive power outputs. At each control point, each DCPC sends the required information to its immediate neighbors and receives the corresponding information from its neighbors.

The total available power of the wind farm is estimated by the distributed estimator which is executed every few seconds. The global reference information including the wind farm power reference and voltage at the POC are synchronized by the distributed finite-time observer. The sensitivity coefficients of bus voltage with respect to power injections are given to each DCPC. For the sensitivity calculation, one option is to use offline power flow analysis and keep constant, which may lead to significant errors. Another option is updating the sensitivity in every control period, whereas it might lead to heavy computation and communication burden. Thus, considering the operation states cannot dramatically change in a short term, the sensitivity coefficients are updated with low frequency in this study. An analytical sensitivity calculation method first developed for radial distribution networks is adopted in this paper [23]. It is expected that the closed-loop nature of MPC will compensate infrequently updated sensitivity coefficients.

The objective of the D-MPC for wind farms includes two parts: 1) active power dispatch; and 2) reactive power/voltage control. For the active power control, the controller minimizes the fatigue loads of WTs while tracking the power reference of the wind farm required by system operators, which is considered as a soft constraint and explicitly expressed in the
objective function. For the reactive power control, the voltages at the POC and WT buses are all taken into consideration. In addition, the fair reactive power sharing is also addressed. To better regulate the voltages, the impact of active power variations on voltages is considered.

3. Predictive Model of a WT

In this section, the incremental state-space models of the WT including active and reactive power control loops are obtained by linearizing the nonlinear WT model at the operating point.

3.1. Active Power Control Loop

Modern WTs can be controlled to operate in the maximum power point tracking (MPPT) mode or a derated mode to track the power reference sent from the DCPC. The power-controlled WT model developed by the National Renewable Energy Laboratory (NREL) is used to represent a variable speed WT for modeling of the active power control loop. The configuration of a power-controlled WT model is illustrated in Fig. 3 [24].
3.1. Modeling

According to Fig. 3 a pitch-controlled WT can be approximately modeled by [8], [24],

\[
\begin{align*}
T_t &= 0.5\pi \rho R^2 \frac{3}{5} C_p (\lambda, \theta) \\
F_t &= 0.5\pi \rho R^2 \frac{3}{5} C_t (\lambda, \theta) \\
\dot{\omega}_t &= \frac{1}{J_t} (T_t - \eta T_g) \\
\dot{\omega}_g &= \eta \dot{\omega}_t \\
J_t &= J_t + \eta^2 J_g \\
T_r &= \frac{\eta^2 J_g T_t + \eta J_t T_g}{J_t} \\
T_g &= \frac{\Delta T_{\text{ref}}}{J_g} \\
\end{align*}
\]

where \( T_t \) and \( F_t \) are fied nonlinear WT model at the operating point, the incremental state-space model can be obtained. Define \( \Delta = \begin{bmatrix} \Delta \omega_g \\ \Delta \omega_t \\ \Delta \theta \\ \Delta \beta \end{bmatrix} \), \( \Delta x_{\text{ref}} = \begin{bmatrix} \Delta \omega_g \\ \Delta \omega_t \\ \Delta \theta \\ \Delta \beta \end{bmatrix} \), and \( \Delta y_{\text{ref}} = \begin{bmatrix} \Delta T_{\text{ref}} \\ \Delta \omega_t \\ \Delta \theta \\ \Delta \beta \end{bmatrix} \).

\[ \begin{align*}
\dot{\Delta} x_p &= A_p \Delta x_p + B_p \Delta u_p + E_p \\
\Delta y_p &= C_p \Delta x_p + D_p \Delta u_p \\
\end{align*} \] (3)

The generator torque and pitch controller can be modeled by [8], [24].

\[ \begin{align*}
T_g &\approx \frac{P_{\text{ref}}}{\mu \dot{\omega}_t} \\
\dot{\omega}_t &= -\frac{\omega_t}{\tau_t} + \frac{1}{\tau_t} \dot{\omega}_g \\
\theta_{\text{ref}} &= \frac{K_p^\theta}{K_0 + K_1 \theta_{\text{ref}}} (\omega_t - \omega_g) + \frac{K_p^\theta}{K_0 + K_1 \theta_{\text{ref}}} \left( \frac{\omega_t - \omega_g}{s} \right) \\
\beta &= \frac{K_p^\beta}{K_0 + K_1 \theta_{\text{ref}}} \theta \\
\end{align*} \] (2)

More details about the pitch-controlled WT can be referred to [8]. The definitions of these variables are presented in the Appendix.

3.1.2. Incremental State-Space Model

The modeling of physical and control systems has been presented above. Based on the linearization of the simplified nonlinear WT model at the operating point, the incremental state-space model can be obtained. Define \( \Delta x_p := \begin{bmatrix} \Delta \omega_g \\ \Delta T_r \\ \Delta F_t \\ \Delta \theta \end{bmatrix} \) is the state variable vector, \( \Delta y_p := \begin{bmatrix} \Delta T_{\text{ref}} \end{bmatrix} \) is the output variable vector, and \( \Delta u_p := \begin{bmatrix} \Delta P_{\text{ref}} \end{bmatrix} \) is the control input. Based on (1) and (2), the linearized incremental state-space model of the active power control loop can be represented as,

\[
\begin{align*}
\Delta x_p &= A_p \Delta x_p + B_p \Delta u_p + E_p \\
\Delta y_p &= C_p \Delta x_p + D_p \Delta u_p
\end{align*}
\] (3)

with

\[
A_p = \begin{bmatrix}
\frac{\eta J_t \partial T_r}{J_t \partial \omega_g} & \frac{\eta^2 P_{g0}}{J_t \omega_{g0}^2} & \frac{\eta J_t \partial \theta}{J_t \partial \beta} \\
\frac{1}{\tau_t} & -\frac{1}{\tau_t} & 0 \\
\frac{-K_p^\theta}{\tau_{\text{ref}}} + K_p^\theta & 0 & 0
\end{bmatrix},
B_p = \begin{bmatrix}
\frac{\eta^2 P_{g0}}{J_t \omega_{g0}^2} \\
0 \\
0
\end{bmatrix},
C_p = \begin{bmatrix}
\frac{\eta^2 J_g \partial T_r}{J_t \partial \omega_g} & \frac{\eta J_t P_{g0}}{\mu J_t \omega_{g0}^2} & \frac{\eta^2 J_g \partial \theta}{\partial \beta}
\end{bmatrix},
D_p = \begin{bmatrix}
\frac{\eta J_t}{\mu J_t \omega_{g0}} \\
0 \\
0
\end{bmatrix},
E_p = \begin{bmatrix}
\frac{\eta J_t (T_{\text{ref}} - \eta T_{g0})}{J_t \omega_{g0}^2} \\
0 \\
0
\end{bmatrix}
\]

where the coefficients \( \frac{\partial T_r}{\partial \omega_g} \), \( \frac{\partial T_r}{\partial \theta} \), \( \frac{\partial F_t}{\partial \omega_g} \) and \( \frac{\partial F_t}{\partial \theta} \) are derived from the Taylor approximation of \( T_r \) and \( F_t \) at the operating point. The discrete model can be obtained by discretizing the continuous state-space model \( 1 \) with the sampling time \( \Delta \tau_s \),

\[
\begin{align*}
\Delta x_p(k+1) &= A_p^d \Delta x_p(k) + B_p^d \Delta u_p(k) + E_p^d \\
\Delta y_p(k) &= C_p^d \Delta x_p(k) + D_p^d \Delta u_p(k) \\
\end{align*}
\] (4)

where \( A_p^d = e^{A_p \Delta \tau_s} \), \( B_p^d = \int_0^{\Delta \tau_s} e^{A_p \tau} B_p d\tau \), \( E_p^d = \int_0^{\Delta \tau_s} e^{A_p \tau} E_p d\tau \), \( C_p^d = C_p \), \( D_p^d = D_p \).
3.2. Reactive Power Control Loop

3.2.1. Modeling

For full-scale converter WTs, the grid side converter can be controlled to provide reactive power support for the grid. A standard control scheme for voltage source converter, consisting of the cascaded control structure (inner loop and outer loop), is implemented in this study. For the active power loop, the dc voltage is controlled at the nominal value. For the q-axis control loop, the reactive power control mode is adopted as illustrated in Fig. 4.

Figure 4: Structure of the reactive power control loop.

By introducing the feedforward terms \( \omega L_C i_d \) and \( \omega L_C i_q \), the \( P \)-control loop and \( Q \)-control loop are decoupled. And by selecting,

\[
K_{\text{inr}}^p = \frac{L_C}{\tau_{\text{inr}}} \quad K_{\text{inr}}^q = \frac{R_C}{\tau_{\text{inr}}}
\]

the whole system can be simplified as Fig. 5.

Figure 5: Simplified equivalent control loop.

Accordingly, the basic formulation of the reactive power control can be expressed as,

\[
\begin{align*}
\dot{i}_q^* &= K_{\text{inr}}^p (Q_{\text{meas}} - Q_{\text{ref}}^W) + K_{i_1}^p \frac{Q_{\text{meas}} - Q_{\text{ref}}^W}{s} \\
\dot{i}_q &= -\frac{1}{\tau_{\text{inr}}} i_q + \frac{1}{\tau_{\text{inr}}} i_q^* \\
Q_{\text{WT}} &= -\frac{3}{2} V_S i_q \\
\dot{Q}_{\text{meas}} &= -\frac{1}{\tau_{\text{m}}} Q_{\text{meas}} + \frac{1}{\tau_{\text{m}}} Q_{\text{WT}}
\end{align*}
\]

where the corresponding variables and parameters are defined in the Nomenclature and Appendix.

3.2.2. Incremental State-Space Model

Introduce an auxiliary state variable \( Q_{\text{int}}^W \) defined as,

\[
Q_{\text{int}}^W := \frac{Q_{\text{meas}} - Q_{\text{ref}}^W}{s}
\]

And define \( \Delta x_q := [\Delta Q_W, \Delta Q_{\text{meas}}, \Delta Q_{\text{int}}^W]^T \) is the state variable vector, \( \Delta y_q := [\Delta Q_W]^T \) is the output variable, and \( \Delta u_q := [\Delta Q_{\text{ref}}^W]^T \) is the control input. Accordingly, the linearized incremental state-space model regarding the reactive power loop can be expressed as,

\[
\begin{align*}
\dot{\Delta x}_q &= A_q \Delta x_q + B_q \Delta u_q \\
\Delta y_q &= C_q \Delta x_q
\end{align*}
\]
with

\[
A_q = \begin{bmatrix}
- \frac{1}{\tau_{inr}} & - \frac{3}{2\tau_{inr}} V_{S0} K_p & - \frac{3}{2\tau_{inr}} V_{S0} K' \n 0 & - \frac{1}{\tau_{m}} & 0 \\
\tau_m & 0 & 1 \\
0 & \tau_m & 1
\end{bmatrix},
\]

(9)

\[
B_q = \begin{bmatrix}
\frac{3}{2\tau_{inr}} V_{S0} K_p \n 0 \\
-1
\end{bmatrix}, C_q = [1 \ 0 \ 0].
\]

(10)

Similarly, the discrete model can be obtained from the continuous state-space model (8) with the sampling time \(\Delta\tau_s\),

\[
\Delta x_q(k + 1) = A_q^d \Delta x_q(k) + B_q^d u_q(k)
\]

\[
\Delta y_q(k) = C_q^d \Delta x_q(k)
\]

where

\[
A_q^d = e^{A_q \Delta\tau_s}, B_q^d = \int_0^{\Delta\tau_s} e^{A_q \tau} B_q d\tau, C_q^d = C_q.
\]

4. D-MPC Based Power Control Design

The proposed power control scheme includes three parts: 1) distributed synchronization (estimation) of the global reference information \(P_{\text{ref}}^\text{WF}\) and \(V_{\text{POC}}\); 2) distributed estimation of available power of the wind farm \(P_{\text{avi}}^\text{WF}\); and 3) MPC formulation of the voltage control problem.

4.1. Synchronization of Global Reference Information

As mentioned in Section 2, the distributed control scheme is developed based on the sparse and connected communication network. The WT can only exchange information with its neighbors and only a few WTs have directly access to the global information. Thus, the global information cannot be broadcast to each WT in centralized manner. In this context, for \(P_{\text{ref}}^\text{WF}\) and \(V_{\text{POC}}\), a distributed finite-time observer can be adopted to estimate them in a distributed way (25), which is,

\[
\hat{P}_{\text{ref}}^\text{WF}(i)(t) = \text{sig} \left( \sum_{j=1}^{N} a_{ij} \left( \hat{P}_{\text{ref}}^\text{WF}(j)(t) - \hat{P}_{\text{ref}}^\text{WF}(i)(t) \right) + g_i \left( P_{\text{ref}}^\text{WF}(t) - \hat{P}_{\text{ref}}^\text{WF}(i)(t) \right) \right)^{\frac{1}{2}}
\]

\[
\hat{V}_{\text{POC}}(i)(t) = \text{sig} \left( \sum_{j=1}^{N} a_{ij} \left( \hat{V}_{\text{POC}}(j)(t) - \hat{V}_{\text{POC}}(i)(t) \right) + g_i \left( V_{\text{POC}}(t) - \hat{V}_{\text{POC}}(i)(t) \right) \right)^{\frac{1}{2}}.
\]

According to the theorem in (25), if the communication network is connected, the estimated value \(\hat{P}_{\text{ref}}^\text{WF}(i)\) and \(\hat{V}_{\text{POC}}(i)\) will synchronize to the global references \(P_{\text{ref}}^\text{WF}\) and \(V_{\text{POC}}\) in a finite time, respectively.

4.2. Distributed Estimation of Total Available Power of the Wind Farm

Similarly, to estimate \(P_{\text{avi}}^\text{WF}\) in a distributed way, the average-consensus protocol can achieve it with the sparse and connected communication network. Thus, the distributed estimator can be designed as (26),

\[
\hat{P}_{\text{avi}}(i)(t) = -\epsilon \sum_{j=1}^{N} a_{ij} \left( \hat{P}_{\text{avi}}(j)(t) - \hat{P}_{\text{avi}}(i)(t) \right)
\]

(12)

where \(\epsilon > 0\) is the constant gain. According to the theorem in (25), if the communication network is connected, the estimated values will converge to a stable equilibrium that is average of the initial values provided by each WT. The estimation process is executed every several seconds. At the beginning of the process, \(\hat{P}_{\text{avi}}(i)(t)\) is initialized with the current available power of WT-i. Similarly, if the communication network is connected, \(\hat{P}_{\text{avi}}(i)(t)\) will definitely converge to \(P_{\text{avi}}^\text{WF}/N\) in a finite time. Accordingly, each WT can estimate \(P_{\text{avi}}^\text{WF}\) by,

\[
\hat{P}_{\text{avi}}^\text{WF} = N \cdot \hat{P}_{\text{avi}}(i).
\]

(13)
4.3. MPC Formulation

In the MPC, the control input is obtained by solving a discrete-time optimal control problem over a given horizon. An optimal control input sequence is produced and only the first control in the sequence is applied. In this paper, the non-cooperative (communication-based) D-MPC is adopted, in which each subsystem (WT) controller anticipates the impact of system interactions only locally and the whole system performance converges to the Nash equilibrium.

For each WT, the control inputs are the active and reactive power commands from the DCPC. In the active power control part, the controller minimizes the fatigue loads represented by fluctuations of low-speed shaft torque $\Delta T_s$ while tracking the power reference $P_{WT}^{ref}$. In the reactive power part, the terminal voltage of the WT and voltage at POC are regulated around 1.0 p.u., which also results in the system active power loss minimization. Besides, the reactive power sharing is also considered. Suppose the prediction and control horizon (steps) are $H_p$ ($N_p = H_p/T_e$) and $H_c$ ($N_c = H_c/T_e$), respectively. As known, $N_p \geq N_c$. From the computational viewpoint, they should be equal, i.e., $N_p = N_c$, unless the controller is required to consider changes beyond the control horizon.

4.3.1. Active Power Control

Supposing $k$ is the current operating point (step), the predictive value and control input of WT-$i$ ($i \in [1, \ldots, N]$) are defined by,

$$\Delta P_{WT}^{ref}(k+l|k) = \Delta u_p^{(i)}(k+l|k),$$

$$\Delta T_s^{(i)}(k+l|k) = S_{T_s} \cdot \Delta y_p^{(i)}(k+l|k),$$

$$\Delta F_i^{(i)}(k+l|k) = S_{F_i} \cdot \Delta y_p^{(i)}(k+l|k),$$

The MPC formulation of WT-$i$ at step $k$ can be formulated as follows:

$$\min_{\Delta u_p^{(i)}(k|k), \ldots, \Delta u_p^{(i)}(k+N_p-1|k)} \sum_{l=1}^{N_p} W_P \left( P_{g}^{(i)} + \Delta P_{WT}^{ref}(k+l|k) - \frac{P_{WT}^{ref}(i)}{N} \right)^2 + \sum_{l=1}^{N_p} W_T \left( \Delta T_s^{(i)}(k+l|k) \right)^2 \quad (14)$$

subject to

$$\Delta P_{WT}^{ref}(k+l+1|k) = A_p^{d(i)} \Delta P_{WT}^{ref}(k|k) + B_p^{d(i)} \Delta u_p^{(i)}(k+l|k) + E_p^{d(i)}(k+l|k),$$

$$\Delta y_p^{(i)}(k+l|k) = C_p^{d(i)} \Delta P_{WT}^{ref}(k|k) + D_p^{d(i)} \Delta u_p^{(i)}(k+l|k),$$

$$0 \leq \Delta P_{WT}^{ref}(i) \leq \Delta u_p^{(i)}(k+l|k) \leq \Delta y_p^{(i)}(k+l|k),$$

where the first term in the cost function is used as the soft constraint of the power reference tracking. The second term and third term are used to penalize the fatigue loads of the WT. $W_P$, $W_T$, and $W_F$ are the weighting factors which penalize for power tracking, shaft torque and thrust force, respectively.

4.3.2. Reactive Power Control

By defining the predicted power output vectors, sensitivity vector and control input vector,

$$\Delta Y_p^{(i)}(k+l|k) = [a_{i1} \Delta P_{WT}^{ref}(i)(k+l|k-1), \ldots, a_{i(i-1)} \Delta P_{WT}^{ref}(i-1)(k+l|k-1), \ldots, a_{iN} \Delta P_{WT}^{ref}(N)(k+l|k-1)]^T,$$

$$\Delta Y_q^{(i)}(k+l|k) = [a_{i1} \Delta y_q^{(i)}(k+l|k-1), \ldots, a_{i(i-1)} \Delta y_q^{(i-1)}(k+l|k-1), \ldots, a_{iN} \Delta y_q^{(N)}(k+l|k-1)]^T,$$

$$S_{V_POC}^{P(i)} = \left[ \frac{\partial V_POC}{P_{W_i}}, \ldots, \frac{\partial V_POC}{P_{W_{i-1}}}, \frac{\partial V_POC}{P_{W_i}}, \frac{\partial V_POC}{P_{W_{i+1}}}, \ldots, \frac{\partial V_POC}{P_{W_N}} \right],$$

$$S_{Q_POC}^{P(i)} = \left[ \frac{\partial V_POC}{Q_{W_i}}, \ldots, \frac{\partial V_POC}{Q_{W_{i-1}}}, \frac{\partial V_POC}{Q_{W_i}}, \frac{\partial V_POC}{Q_{W_{i+1}}}, \ldots, \frac{\partial V_POC}{Q_{W_N}} \right],$$

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The available power estimation is executed every 10 s and the sensitivity coefficients are updated every 120 s. The control period of the D-MPC is set as 1 s and $N_p = N_q = 5$. The simulation results of the proposed distributed control
scheme are compared with the conventional centralized control. For the active power control, the proportional dispatch strategy is used according to the available power is given as,

\[ P_{WT}^{ref}(i) = P_{WF}^{ref} \times P_{WT}^{avi}(i) \] \hspace{1cm} (22)

For reactive power control, the conventional centralized optimal voltage control [12] is adopted, which solves the following optimization problem at each control step,

\[
\begin{align*}
\min_{Q_{W1}, \ldots, Q_{WN}} & \quad W_{POC}(V_{POC} - V_{ref})^2 + \sum_i W_W (V_{Wi} - V_{ref})^2 + W_Q \sum_{i,j} \left( \frac{Q_{Wi}}{\partial_{W_i}} - \frac{Q_{Wj}}{\partial_{W_j}} \right)^2 \\
\text{subject to} & \quad Q_{W} \leq Q_W \leq \bar{Q}_W \\
& \quad V_{POC} = V_{POC0} + \frac{\partial V_{POC}}{\partial Q_{W1}} \Delta Q_{W1} + \cdots + \frac{\partial V_{POC}}{\partial Q_{WN}} \Delta Q_{WN}, \\
& \quad V_{W} = V_{W0} + \frac{\partial V_{W}}{\partial Q_{W1}} \Delta Q_{W1} + \cdots + \frac{\partial V_{W}}{\partial Q_{WN}} \Delta Q_{WN}.
\end{align*}
\] \hspace{1cm} (23)

The sampling time of the centralized controller is designed as 1 s.

5.1. Performance of the Estimator and Observer

The total simulation time is 600 s. The power reference and available power of the wind farm are shown in Fig. 9. The power demand changes from 20 MW to 30 MW at \( t = 200 \) s and from 30 MW to 25 MW at \( t = 470 \) s.
The performances of the estimator for the total available power of the wind farm and the observer for wind farm power reference are presented in this subsection. The estimated available power of the wind farm is shown in Fig. 10. It can be observed that all the estimators can fast converge to the actual value within about 1 s after every updated point (every 10 s). The synchronization of the wind farm power reference is shown in Fig. 11. As can be seen, the distributed observers can fast track the variations of the power reference, implying the global information can be effectively synchronized in the wind farm.

5.2. Power Tracking and Fatigue Loads Reduction
5.2.1. Power Tracking

To avoid sharp power variations of WTs, the (estimated) wind farm power reference is sent to the distributed controller (after the observer) or central controller through a first-order low-pass filter with the time constant of 10 s. The power reference tracking performance of the proposed scheme is illustrated in Fig. 12. It can be seen that the centralized
controller can precisely track the power reference of the wind farm. The D-MPC can also regulate the output power within small deviations, implying the power tracking performance of the proposed scheme is acceptable.

### 5.2.2. Fatigue Loads Reduction

WT-04 is used as an example to validate the fatigue loads reduction performance. The shaft torque and thrust force of WT-04 are shown in Fig. 13 and Fig. 14 respectively. Compared with the centralized control, the variations of $T_s$ and $F_t$ are reduced with D-MPC, implying the less fatigue loads experienced by WTs. The standard deviations $T_s$ and $F_t$ for all WTs are listed in Table II. For $T_s$, the standard deviations are reduced by 5%–8% and for $F_t$, they are 27%–59%.
5.3. Voltage Regulation

In this subsection, the voltage control performance is evaluated. According to experimental studies in [27], the minimal information update interval in IEEE 802.11 is in the order of 10 ms. For the commonly used fiber optical, the time delay might be in the range of several milliseconds (less than 1.5 ms for 40 km) [28]. They both can be used for distributed control. Thus, to further illustrate the potential benefits of D-MPC for voltage control, the sampling time of the D-MPC is reduced to 0.2 s. The voltage reference $V_{\text{ref}}$ is set as 1.0 p.u.. A reactive power load is suddenly added to Bus06 at $t = 120$ s and removed at $t = 420$ s. Fig. 15 shows the voltage at the POC. WT-04 (closest bus along the feeder) and WT-08 (furthest bus along the feeder) are selected as the representative WTs. Their terminal voltages are shown in Figs. 16 and 17.
As shown in Fig. 15, both the centralized control and D-MPC can efficiently regulate the $V_{POC}$ within the feasible range of $0.99 \sim 1.01$ p.u., implying the good voltage regulation performance. After the disturbances at $t = 120$ s, $V_{POC}$ suddenly falls to 0.99 p.u. and then it recovers to 0.995 p.u. due to the reactive power support of WTs. After the disturbance at $t = 420$ s, $V_{POC}$ suddenly exceeds 1.004 p.u., and then it recovers close to 1.0 p.u.. And as can be seen, the overall performances of the D-MPC and the centralized control, which have the same control period 1 s, are similar. However, if the control period of the D-MPC can be shortened to 0.2 s, the D-MPC can significantly improve the voltage recovery performance, which shows the advantages of the D-MPC. Similarly, as can be seen from Figs. 16 and 17 the terminal voltages of WTs, including the furthest WT along the feeder, i.e. WT-08, can be effectively controlled around 1.0 p.u.. Compared with the centralized control, the D-MPC with shorter control period shows better control performance.
5.4. Comparison of the Coordinated and Separated Control

In this subsection, the comparison of the coordinated (proposed in this paper) and the conventional separated active power and reactive control of a wind farm is presented. The simulation time is 100 s. It can be seen from Figs. 18 and 19 that both the coordinated D-MPC and separated D-MPC can effectively regulate the voltage within the feasible range. Compared with the separated D-MPC, the coordinated D-MPC can better control the voltage with smaller deviations and fluctuations, which further demonstrates the advantages of the proposed control scheme.

6. Conclusion

This paper proposes a distributed coordinated active and reactive power control scheme for wind farms based on the MPC, in which the distributed reactive/voltage control of wind farms is achieved and coordinated with the active power control. The simulation results validate that the proposed control scheme can effectively reduce the fatigue loads experienced by WTs while tracking the wind farm power reference. And it can regulate the voltages within the feasible
range. By shortening the control period, the control performance of the D-MPC can be significantly improved and better than the centralized control. Compared with the separated active and reactive power control, the proposed coordinated control can better regulate the voltages with smaller fluctuations and deviations by taking into account the impact of active power variations on voltages. Besides, since the distributed control scheme is center-free and only requires sparse communication network, it could be helpful to significantly reduce the costs for large-scale wind farms, implying better scalability and economic performances.

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Appendix

The parameters of WTds and the wind farm are presented in the Tables 2 and 3.

Table 2
Parameters of wind farm collector system.

<table>
<thead>
<tr>
<th>Unit</th>
<th>Parameters</th>
</tr>
</thead>
<tbody>
<tr>
<td>33kV Cable</td>
<td>$R = 0.0975 \Omega/km$, $L = 0.38 \mu H/km$, $C = 0.24 \mu F/km$</td>
</tr>
<tr>
<td>33kV/110kV Transformer</td>
<td>$S_N = 50MVA$, $R = 0.002$ p.u., $X = 0.12$ p.u.</td>
</tr>
</tbody>
</table>

Table 3
Parameters of a 5-MW wind turbine model.

<table>
<thead>
<tr>
<th>Symbol</th>
<th>Description</th>
<th>Value</th>
</tr>
</thead>
<tbody>
<tr>
<td>$\eta$</td>
<td>Gearbox ratio</td>
<td>97</td>
</tr>
<tr>
<td>$R$</td>
<td>Rotor radius [m]</td>
<td>63</td>
</tr>
<tr>
<td>$J_l$</td>
<td>Rotor inertia [kg-m(^2)]</td>
<td>3.544 \times 10^7</td>
</tr>
<tr>
<td>$J_g$</td>
<td>Generator inertia [kg-m(^2)]</td>
<td>534.116</td>
</tr>
<tr>
<td>$\omega_0$</td>
<td>Rated generator speed [rad/s]</td>
<td>122.9096</td>
</tr>
<tr>
<td>$\mu$</td>
<td>Generator efficiency</td>
<td>0.944</td>
</tr>
<tr>
<td>$\rho$</td>
<td>Air density</td>
<td>1.2231</td>
</tr>
<tr>
<td>$K_p$</td>
<td>Proportional gain of the pitch control</td>
<td>0.1965</td>
</tr>
<tr>
<td>$K_i$</td>
<td>Integral gain of the pitch control</td>
<td>0.0842</td>
</tr>
<tr>
<td>$K_0$</td>
<td>Gain scheduling coefficient</td>
<td>1</td>
</tr>
<tr>
<td>$K_1$</td>
<td>Gain scheduling coefficient</td>
<td>1.9412</td>
</tr>
<tr>
<td>$K_{out}$</td>
<td>Proportional gain of the converter control</td>
<td>25</td>
</tr>
<tr>
<td>$K_{out}$</td>
<td>Integral gain of the converter control</td>
<td>5000</td>
</tr>
<tr>
<td>$\tau_m$</td>
<td>time constant of low-pass filter [ms]</td>
<td>10</td>
</tr>
<tr>
<td>$\tau_{int}$</td>
<td>time constant of low-pass filter [ms]</td>
<td>5</td>
</tr>
</tbody>
</table>

References


