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Digital Non-Linear Equalization for Flexible Capacity Ultradense WDM Channels for Metro Core Networking

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Abstract: We experimentally demonstrate that digital non-linear equalization allows for using independent tunable DFB lasers spaced at 12.5 GHz for ultradense WDM PM-QPSK flexible capacity channels for metro core networking.

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1. Introduction
Rigid and coarse granularity of wavelength-division-multiplexing (WDM) systems leads to inefficient capacity utilization. In flexible optical WDM (FWDM), spectral resources are allocated in a flexible and dynamic way; channel spacing and center wavelength are not fixed on the ITU-T grid, resulting in higher spectral efficiency [1-2]. Even in this higher-efficient bandwidth allocation scenario, the bandwidth is non optimally utilized as large guard bands are still employed. Ultradees (UD) WDM with a channel spacing of less than 25 GHz, may provide an evolutionary path from conventional infrastructures towards more scalable and spectrally efficient networks. This trend is supported by the increased demand for more capacity, flexibility and upgradeability of transmission technique while keeping some legacy ones, in particular in the metro segments of the optical network. Increasing the number of wavelengths within a fixed optical bandwidth (e.g., C band), by decreasing the spacing between neighboring channels, allows an increase in the system capacity without requiring of high-speed electronics (e.g., >40 Gb/s), while keeping compatibility with the 10 Gb/s SONET/SDH equipment. UDWM systems with no aliasing condition (i.e. with channel spacing higher than the double the baud rate) have been studied, with particular attention to limitations introduced by fiber nonlinearity effects such as Four-Wave Mixing (FWM), Cross-Phase Modulation (XPM), fiber chromatic dispersion-induced symbol intersymbol interference (ISI) [3]. However, to cope with the required bandwidth efficiency, future UDWM schemes will need to use extremely close channel spacing; this implies taking measures to mitigate the resultant detrimental effects from crosstalk and neighboring channels interferences.

In this paper we experimentally demonstrate that by employing a digital nonlinear equalizer, such as a Decision Feedback Equalizer (DFE), we can mitigate inter-channel interference, improve overall system performance in terms of OSNR and allow for a channel spacing of 12.5 GHz using conventional independent DFB light sources for a 40 Gb/s ultradees 3 channel WDM PM-QPSK system with coherent detection. Our proof of principle experiment demonstrates that in a 50GHz bandwidth (in accordance to the ITU-T grid) up to 4 channels can be transmitted, improving the total bit rate from 40 Gb/s to 160 Gb/s per slot, with small changes in the electronic equipment.

2. System setup
The general outline of the experimental setup for the UDWM polarization multiplexing (PM) QPSK coherent optical (CO) system is shown in Fig. 1. At the transmitter side three carriers are generated employing three independent tunable distributed feedback lasers (DFB) with 10 MHz linewidth; one of them is fixed at a central wavelength ($\lambda_c$ of 1550.511 nm. Different channel spacing values $\Delta \lambda$ (20, 18, 16, 14, 13, 12, 11 and 10 GHz) between the 3 carriers have been realized by changing the wavelength of the right ($\lambda_d$) and left ($\lambda_l$) DFB lasers as to have the desired spectral separation. A 40 GHz bandwidth photodiode and an Electrical Spectrum Analyzer are used to verify correct spacing between the three channels. A polarization beam splitter divides the signal into two orthogonal polarization which are then fed into two optical I/Q modulator (nested Mach-Zender modulator). A 10 Gb/s pattern generator (PPG) generates the pseudo random binary sequence (PRBS), with $2^{15}$-1 bit length, that drives the two QPSK modulators. Two uncorrelated branches of polarization orthogonal QPSK signals are then combined with a polarization beam combiner (PBC) to generate the PM QPSK signals, at 10 Gbaud/s. An 80 km span of standard single mode fiber (SMF) is used as optical transmission link. At the receiver side an optical tunable band-pass filter (0.33 nm or 37.5 GHz full width at half maximum, FWHM, at1550 nm) is employed before the optical pre-amplifier; a second band-pass filter (0.5 nm or 62.5 GHz FWHM at 1550 nm) rejects the out of band ASE noise. An external cavity laser (ECL) with 100 kHz linewidth is used as local oscillator (LO). The PM
coherent receiver consists of two 90° hybrids and balanced photodetectors. The photodetected inphase and quadrature outputs are sampled at 40 GS/s for offline demodulation. Digital signal processing (DSP) algorithms implement digital filtering, PM QPSK constant modulus algorithm (CMA) equalization, QPSK carrier phase recovery and bit error decision.

As the channel distortion or the ISI of a transmission system is too severe for a linear equalizer to mitigate the channel impairments, non-linear equalizer has been used. A DFE is a non-linear structure that uses previous detector decision to eliminate the ISI on pulses currently demodulated. The DFE equalizer will not amplify the noise, cause according to its structure, the equalization process is done through the feedback, noiseless, data. A nonlinear DFE consisting of a feedforward filter (FFE) and a feedback filter (FBE) is used in our DSP receiver, after the carrier phase recovery block, to improve the system performances [4]. The taps of the two equalizers are adjusted using a least mean square (LMS) stochastic algorithm.

3. Results

After optimize the PM CMA algorithm structure, we've investigated the impact of a nonlinear decision feedback equalizer consisting of a 1 tap feedforward filter (FFE) and a 7 or 9 taps feedback filter (FBE) on the system performances (this structure is indicated as DFE in the graphs); the digital filter is then re-optimized to improve further the BER curves. The measured BER performances of the UDWDM PM QPSK for back-to-back (B2B) system are shown in Fig. 2. Fig. 2(a) shows the BER experimental performances as a function of the measured OSNR for a spacing of 14 GHz without nonlinear equalization, with the nonlinear equalization structure DFE and with further optimization of the digital filter. It can be observed that the non-linear equalization and the optimization

![Figure 1: Experiment setup of UDWDM PM QPSK system; DFB: Distributed feedback laser; PD: 40G photodiode; ESA: Electrical Spectrum Analyzer; PC: polarization controller; PBS: polarization beam splitter; PBC: polarization beam combiner; EDFA: erbium-doped fiber amplifier; VOA: variable optical attenuator; OBPF: optical band-pass filter; ECL: external cavity laser.](image1)

![Figure 2: (a) BER as a function of OSNR for a spacing of 14GHz without nonlinear equalization, with DFE and with optimization of the digital filter. (b) BER as a function of the spacing for two fixed values of OSNR with and without nonlinear equalization.](image2)
of the digital filter afterward, can improve the system performances up to 4.5 dB in terms of OSNR and 0.6 dB in terms of BER. The BER versus carrier spacing for two fixed value of OSNR is shown if Fig. 2(b); the results show that the DSP implementation can improve the experimental BER results for all the different spacing. For a fix spacing the algorithm can improve the BER result, while for a fix BER value the use of DSP, it allows closer channel. Of particular importance is 12.5 GHz of spacing which, thanks to the nonlinear equalization, shows performances better then the UFEC limit for both the chosen OSNR.

Fig. 3 shows OSNR performance as a function of the spacing with and without nonlinear equalization for Back-to-Back and 80 km of SMF optical transmission are presented. The BER is fixed @10^{-4} and @10^{-3}, both below the UFEC level. The improvement enabled by the DSP implementation on both BER and spacing values is remarkable. As shown in Fig. 3(a) we can improve ~2-3 dB for OSNR with the same spacing value between the carriers; on the other hand for a fix OSNR the carriers can be generated 6 GHz closer, moving from 20 GHz of spacing, case where we have no aliasing, to 14 GHz of spacing. The same behavior is obtained for 80 km of optical transmission. It can be observed in Fig. 3(b) that with the same spacing value between the carriers, the improvement in terms of OSNR is ~3 dB for BER@10^{-3} and ~5dB for BER@ 10^{-4}; for the same OSNR then the carriers can be ~6 GHz closer, moving to ~19 GHz to 13 GHz. It should be noticed that the length of the FBE is 7 or 9 taps compared to 1 tap for the FFE, indicating that the signal is affected by non-linearities.

The results obtained suggest more efficient and flexible utilization of the available bandwidth. In case of ITU-T grid with 50 GHz spacing, one single 40 Gb/s signal can be transmitted per slot. Using the proposed UDWDM with DSP non linear equalization, up to 4 channels (12.5 GHz of spacing) per slot can be transmitted with a total bit-rate of 160 Gb/s, with small changes in the electronic equipment. Better results per slot could also be obtained with a single carrier higher data rate signal (100 Gb/s), but this would imply higher speed electronics and no more compatibility with the existing 10 Gb/s SONET/SDH equipment.

4. Conclusion

We have experimentally demonstrated that by employing a DFE nonlinear equalizer in the DSP domain, an ultradense WDM PM QPSK can be realized. Performances below the UFEC limit are obtained for very closed channel spacing (down to 12.5 GHz). In a 50 GHz ITU-T grid, the structure proposed allows to quadruple the number of user in a flexible way, moving from a total bit rate of 40 Gb/s per slot to 160 Gb/s.

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6. References