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Wang, Yun; Gong, Wenming; Wu, Qiuwei

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Robust Current Control of Doubly Fed Wind Turbine Generator under Unbalanced Grid Voltage Conditions

Yun Wang  
College of Mechatronics and Control Engineering, Shenzhen University  
Shenzhen, P.R.China  
wangyun@szu.edu.cn

Wenming Gong  
University of Chinese Academy of Sciences, Institute of Electrical Engineering, Chinese Academy of Sciences, Beijing, P.R.China

Qiuwei Wu  
Centre for Electric Power and Energy (CEE), Department of Electrical Engineering, Technical University of Denmark

Abstract—This paper presents the design of a $H_\infty$ current controller for doubly fed induction generators (DFIGs) in order to maintain stable operation under unbalanced voltage conditions. The $H_\infty$ current controller has a multi-input and multi-output (MIMO) structure and is designed using the loop shaping method. Case studies have been carried out in order to verify the efficacy of the proposed $H_\infty$ current controller for DFIGs. The case study results show that the proposed $H_\infty$ current controller can realize different control objectives, i.e. stable stator current, stable stator active power and stable stator reactive power. It is also shown that the $H_\infty$ current controller is less sensitive to the parameter perturbation.

Index Terms—DFIG, $H_\infty$ current controller, MIMO, unbalanced voltage conditions

I. INTRODUCTION

Unbalanced voltage conditions in power systems are one of the frequent fault conditions during operations [1], [2]. Because the stator of doubly fed induction generators (DFIGs) is directly connected to the grid and the negative sequence impedance of DFIGs is usually small, the unbalanced voltage conditions will cause current unbalance, armature overheating, torque and power pulsation, etc. and will impact the secure and stable operation of DFIGs [2]-[4].

The control methods for DFIGs under unbalanced voltage conditions can be divided into two groups: (1) obtain the positive and negative sequence components of voltages and currents by positive and negative sequence separation, and implement closed loop current control using PI controllers in the positive and negative synchronous rotating frames [3]-[7]. Although the theory of such control methods is simple, the control system is complex and consists of a number of PI controllers. (2) replace PI controllers with more advanced controllers in order to realize better AC current pulsation following characteristics for rotor currents. Among the advanced controllers, the proportional resonant (PR) control or proportional integral resonant (PIR) controller is the most commonly used one [8], [9]. However, the performance of the PR controller depends on the resonant point parameters and the parameters should be adjusted according to the machine speed change, therefore, the applications of such methods are limited.

In practice, the operating conditions of wind turbine generators (WTGs) is very complex and the existing methods have not considered the impact of machine parameter change and disturbance uncertainties on the performance of the complex control system. Since the $H_\infty$ control has successful applications when there are uncertainties of model parameters and disturbances, the robust control method is used to develop a control scheme for DFIGs under unbalanced voltages in order to handle the AC pulsation disturbances. The proposed robust current controller for DFIGs is developed by using the loop shaping method to efficiently suppress the AC pulsation under unbalanced voltage condition. The case study results show that the proposed robust current controller has better robustness and reliability compared to the PIR controller.

II. MATHEMATICAL MODEL OF DFIGS

In the synchronous rotating reference frame, with the rotor side voltage and current converted to the stator side, the equivalent circuit of DFIGs is shown in Figure 1.

![Figure 1 Equivalent circuit model of DFIG](image)

The mathematical model of DFIGs is represented by (1) and (2). Since there is no neutral connection for the DFIG grid connection and hence no zero sequence loop, any vector of DFIGs can be decomposed into positive sequence and negative sequence components.
\[
\begin{align*}
\psi_e &= l_a I_a + l_s I_r, \\
\psi_i &= l_n I_n + l_r I_r, \\
V_s &= r_s I_s + \frac{1}{\omega_s} \frac{d\psi_s}{dt} + j\omega_s \psi_s, \\
V_r &= r_r I_r + \frac{1}{\omega_r} \frac{d\psi_r}{dt} + j\omega_r \psi_r,
\end{align*}
\]
\[
\omega_{\text{up}} = \omega_s - \omega_r
\]

According to the symmetric component method, any vector can be decomposed into positive and negative sequence components and projected to the positive and negative sequence synchronous rotating coordinate reference which is shown in Figure 2.

![Figure 2 projections of vector F on positive and negative sequence synchronous rotating coordinate reference frames](image)

Because the wind turbine generators (WTGs) and the grid connection circuits are usually three phase without neutral, there is no zero sequence loop. The DFIG model can be decomposed into positive and negative sequence components. In the decomposed model, the superscripts ‘+’ and ‘−’ denote the positive sequence and negative sequence rotating reference frame, and the subscripts ‘+’ and ‘−’ denotes the positive and negative sequence components. According to [13], the control commands of negative sequence current can be divided into three categories:

1) To realize balanced rotor current in order to avoid unbalanced heating in the rotor windings.

\[
\begin{align*}
I^+_{rd} &= 0, \\
I^-_{rd} &= 0
\end{align*}
\]

2) To realize stable stator active power output in order to ensure stable active power output from wind power plants.

\[
\begin{align*}
I^+_{iq} &= \frac{V^+_{qr}}{V^+_{qr}} I^+_{rd} + \frac{V^-_{qr}}{\omega_L} I^+_{iq} - \frac{2V^-_{qr}}{\omega_L} I^+_{iq}, \\
I^-_{iq} &= \frac{2V^-_{qr}}{\omega_L} I^+_{iq} + \frac{V^-_{qr}}{\omega_L} I^-_{iq} + \frac{V^+_{qr}}{\omega_L} I^-_{iq}
\end{align*}
\]

3) To realize stable stator reactive power output

\[
I^+_{ad} = \frac{V^+_{qr} I^+_{rd} + V^-_{qr} I^+_{iq}}{V^+_{ad}}, \\
I^-_{ad} = \frac{V^-_{qr} I^-_{iq} + V^+_{qr} I^-_{ad}}{V^+_{ad}}
\]

III. DESIGN OF $H_\infty$ CURRENT CONTROLLER FOR DFIGs

A. Design of $H_\infty$ Controller using the Loop Shaping Method

The objective of the $H_\infty$ controller is to efficiently track the positive and negative sequence components of the machine rotor current and maintain the stability of the control system under certain types of disturbances.

According to the $H_\infty$ control formulation, (1) and (2) can be rewritten as (7) – (11).

\[
G: \begin{cases}
\dot{x} = Ax + Bd + Bu \\
y = Cx
\end{cases}
\]

where \(x\) is the state variable, \(d\) is the disturbance, \(u\) is the control variable, \(y\) is the measured value of the feedback.

![Figure 3. Diagram of DFIG $H_\infty$ current controller](image)

B. Parameter design for the $H_\infty$ controller

For the hybrid $H_\infty$ loop shaping issue, the design of the controller can be converted to find a stable controller $K$ such that the closed loop transfer function of the generalized control object, $P$, can meet the requirement that the $H_\infty$ norm is minimized, i.e. by defining a proper peak value margin $\gamma$ to ensure,
\[
\begin{bmatrix}
W_{p}S_{p}K & -W_{p}SKG_{y} \\
W_{p}S_{o} & -W_{p}SG_{y}
\end{bmatrix} < \gamma
\]  
(12)

where \( S_{p} = (I + KG) \) and \( S_{o} = (I + G_{y}K) \) are input and output sensitivity functions.

(12) shows that the performance of the system mainly depends on the shape of \( ||1/W_{p}||_{\infty} \) and \( ||1/W_{u}||_{\infty} \). Because the order of the \( H_{\infty} \) controller is the sum of the order of the control object and the weight function, in order to reduce the difficulty of using the \( H_{\infty} \) controller, it is important to select a weight function which can meet the control requirements and is with a low order.

After the weight function is selected, the optimal or sub-optimal solution of the controller can be obtained by solving the Algebraic Riccati equations (the iterative calculation is required to obtain the solution). Because the per unit model is used, \( W_{u} \) is set to 1. In order to realize the control objective, a complex function weight function is defined,

\[
W = \frac{s}{M + \omega_{l}} + \frac{k\omega_{l}^{2}}{s^2 + 2\zeta\omega_{l}s + \omega_{l}^2}
\]  
(13)

Where the first part is for the basic control performance, the second part is to determine the control performance at the specified resonant frequency. \( \omega_{l} \) (100Hz) is the low crossover frequency, due to the second part of the weight function, the bandwidth of the controller is larger than the crossover frequency. It is good to constraint the rotor mode by increasing the bandwidth of the current controller, however, it will reduce the damping ratio of the stator mode. When the bandwidth of the current loop is approaching the resonance frequency, the system sensitivity will drastically increase at the stator mode frequency and the system capability against disturbances decrease. Therefore, it is better to set the current control bandwidth below 10dB. In this paper, \( \omega_{l} = 100Hz \), \( \zeta = 0.02 \), \( k = 3 \).

C. Robust stability analysis

When there are uncertainties of system parameters, the robust stability of a system can be quantified by calculating the structural singular value \( \mu \),

\[
\mu = \frac{1}{\min\{\delta(\Delta) : \det(I-M\Delta) = 0\}}
\]  
(4)

When \( \mu \) is less than 1, the system is stable under all parameter perturbation. The smaller \( \mu \) is, the better the system robustness is. According to the control bandwidth requirement explained in the previous section, a \( H_{\infty} \) controller (K_Hinf) and a PIR controller have been designed and the singular value of the two controllers have been calculated which is shown in Figure. It is shown that the \( H_{\infty} \) controller is less sensitive to the parameter perturbation than the PIR controller.

IV. CASE STUDIES

In order to verify the efficacy of the proposed \( H_{\infty} \) controller, case studies have been carried out using MATLAB.

<table>
<thead>
<tr>
<th>TABLE I.</th>
<th>SIMULATION PROCESS DESCRIPTION</th>
</tr>
</thead>
<tbody>
<tr>
<td>t(s)</td>
<td>5-5.2</td>
</tr>
<tr>
<td>Control</td>
<td>Balanced</td>
</tr>
<tr>
<td>objectives</td>
<td>Rotor current</td>
</tr>
</tbody>
</table>

<table>
<thead>
<tr>
<th>TABLE II.</th>
<th>DFIG PARAMETERS</th>
</tr>
</thead>
<tbody>
<tr>
<td>Stator</td>
<td>Resistance</td>
</tr>
<tr>
<td>Inductance</td>
<td>3.6710*(1±2%)</td>
</tr>
<tr>
<td>Mutual</td>
<td>inductance</td>
</tr>
</tbody>
</table>

The case study results are shown in Figure 5. The subplots are stator three phase voltage, stator dq axis currents, rotor dq axis currents, and stator active and reactive power from the top to the bottom. During the time period of \( t=5-5.2 \) s, the stator dq axis current waveforms do not have AC pulsation and are balanced, during the time period of \( t=5.2-5.3 \) s, by injecting certain stator negative current, the stator active power is stable, during the time period of \( t=5.3-5.4 \) s, the reactive power is stable. The case study results show that the \( H_{\infty} \) controller can efficiently control the positive and negative components of the rotor current and realize the control objectives.
V. CONCLUSION

Under unbalanced voltage conditions, the rotor voltages and currents of DFIGs have negative sequence component. In order to keep stable operation of generators, it is essential to properly control the negative sequence component to realize different control objectives. An $H^\infty$ current controller for DFIGs is proposed in this paper and the efficacy of the proposed $H^\infty$ current controller has been verified by simulations.

The simulation results show that the proposed $H^\infty$ current controller has the following advantages.

1) The control system of DFIGs is simplified by replacing the conventional PI controller of the current control loop with the proposed $H^\infty$ current controller.

2) In the synchronous rotating frame, the proposed $H^\infty$ current controller can realize multi-objective control and has better performance.

3) The comparison of the structure singular values shows that the proposed $H^\infty$ current controller has better robustness than the conventional PI controller and is less sensitive to the external disturbances or internal machine parameters.

REFERENCES:


