All-optical ultra-high-speed OFDM to Nyquist-WDM conversion

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Abstract: We propose an all-optical ultra-high-speed OFDM to Nyquist-WDM conversion scheme based on complete OFT. An 8-subcarrier 640 Gbit/s DPSK OFDM super-channel is converted to eight 80-Gbit/s Nyquist-WDM channels with BER<10^-9 performance for all channels. 

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1. Introduction

Due to the rapid traffic growth in optical communication networks, intensive efforts have been made to use the available bandwidth of optical fibers more efficiently. Recent studies have focused on orthogonal frequency division multiplexing (OFDM) and Nyquist wavelength division multiplexing (Nyquist-WDM), due to their high spectral efficiency (SE), which enable a channel spacing equal to the symbol rate. These multiplexing techniques have been used to demonstrate high SE super-channels with Tbit/s capacity [1-2]. The capability to switch between ultra-dense OFDM and Nyquist-WDM networks could become a very important functionality for the next generation communication systems. However, such functionality is presently not possible unless complex optical/electrical/optical conversion is used. Over the last ten years, time-domain optical Fourier transformations (OFT), also known as optical time lenses, have been shown to be a very versatile and powerful tool for ultrafast signal processing. It can be used for manipulating the shapes of optical signals in both the time- and frequency-domains. This has led to demonstrations of e.g. distortion-less transmission [3], optical time-division multiplexing (OTDM) to WDM conversion [4], all-optical OFDM transmission systems [5], and most recently WDM to Nyquist-OTDM conversion using a complete OFT [6].

In this paper, we propose an all-optical ultra-high speed OFDM to Nyquist-WDM conversion scheme based on complete OFT. Using this scheme, we simultaneously convert an 8-subcarrier 640 Gbit/s differential phase-shift keying (DPSK) OFDM super-channel to eight 80-Gbit/s Nyquist-WDM channels. The total signal bandwidth of 800 GHz remains unchanged after conversion, thus maintaining the SE at 0.8 symbol/s/Hz. A full bit error rate (BER) characterization is performed, demonstrating BER<10^-9 performance for all converted channels. To the best of our knowledge, this is the first demonstration of OFDM to Nyquist-WDM conversion.

2. Principle

The OFDM signal is composed of sinc-shaped subcarrier spectra with rectangular waveforms. In contrast, the Nyquist-WDM signal is composed of WDM channels with rectangular spectra and sinc-shaped waveforms. Thus, the OFDM to Nyquist-WDM conversion can be realized by exchanging the temporal and spectral profiles, which can be done by a complete OFT [6]. The principle of the proposed scheme is shown in Fig. 1, including also an all-optical OFDM transmitter and a Nyquist-WDM receiver. In the all-optical OFDM transmitter, a multi-carrier transmitter generates a number of channels with frequency spacing Δf0 at symbol rate fs (A), where Δf0 is N times fs. The multi-carrier channels with flat-top temporal waveforms are bit-wise synchronized. A rectangular temporal gate with width Tps =1/Δf0 is then used to select the center of the overlapping waveforms, resulting in sinc-shaped multi-carrier channels in the frequency-domain with a half zero-crossing width (HZCW) equal to Δf0 (B). The signal is
then OTDM-multiplexed by a factor N, resulting in an OFDM signal with a symbol rate equal to the channel spacing (C), where \( N_f = \Delta \omega_1 \). The complete OFT for OFDM to Nyquist-WDM conversion is based on two quadratic phase-modulation stages (time lenses) with chirp rate \( K \), separated by a dispersion medium of \( D = \beta_2 L \) (where \( \beta_2 \) is 2nd order dispersion and \( L \) is the length), which satisfy the condition \( K = 1/D \) (a K-D-K configuration). This configuration enables both time-to-frequency and frequency-to-time conversions at the same time, thus performing an exchange between the temporal and spectral profiles of the input signal (a “complete” OFT). The chirp rate \( K \) determines the scaling factor between the time and frequency domains according to \( \Delta t = 2\pi \Delta \omega / K \). After the OFT, the OFDM signal is converted to a Nyquist-WDM signal with channel spacing \( \Delta \omega_2 \) depending on the choice of \( K \) (D). To achieve the maximum SE after conversion, the chirp rate \( K \) should be set equal to \( 2\pi \Delta \omega_2 \). However, since the practical OFT process requires a short guard interval (GI) in each OFDM symbol slot for the transition of the quadratic phase-modulation, the \( K \) is usually set to a value less than \( 2\pi \Delta \omega_2 \). Note that the proposed scheme enables OFDM to Nyquist-WDM conversion at the super-channel level (not at the subcarrier-level), in the sense that an individual Nyquist-WDM channel is converted from the corresponding OFDM temporal tributary. In the receiver, the converted Nyquist-WDM super-channel is first WDM-demultiplexed into individual Nyquist-WDM channels (E) using a rectangular optical band-pass filter (OBPF) with bandwidth equal to the channel spacing \( \Delta \omega_2 \). The demultiplexed Nyquist-WDM channel is subsequently sampled at the inter-symbol-interference (ISI)-free point using a narrow optical sampling gate [7]. Finally, the sampled signal (F) is detected by a base-rate receiver. As the OFT is transparent to the data-format [8], this scheme could potentially be applied to advanced modulation formats.

3. Experimental setup and results

The experimental setup is shown in Fig. 2. Eight distributed feedback laser diodes (DFB-LDs), centered from 1547.2 nm to 1552.8 nm with 100 GHz spacing, are used in the transmitter. The outputs of the CW lasers are DPSK-modulated with a 10 Gbit/s 2^31-1 PRBS in a Mach-Zehnder modulator. A 10-ps rectangular temporal gate is used to simultaneously shape all multi-carrier channels into 100 GHz HZCW sinc-shaped multi-carrier channels. The temporal gate is implemented by a non-linear polarization-rotating loop (NPRL) [9] using a 10-ps rectangular control pulse. The obtained eight 10-Gbaud sinc-shaped multi-carrier channels are then OTDM-multiplexed to 80-Gbaud using an OTDM multiplexer, resulting in an 8-subcarrier 640 Gbit/s OFDM super-channel with SE at 0.8 symbol/s/Hz, as shown in Fig. 3 (a). Fig. 3 (a) also shows the 8 OFDM subcarriers with 100 GHz spacing obtained from the individual CW lasers, where the characteristic sinc-shape can be observed. The resulting OFDM waveform is shown in Fig. 3(b), where the 80-Gbaud rectangular-shaped OFDM waveform is obtained. The bit slot is 11.5 ps including a 1.5-ps GI, and an extra 8-ps GI between every 8 tributaries for OFT operation.

A complete OFT is used for the OFDM to Nyquist-WDM conversion. The quadratic phase modulation is implemented based on a four-wave mixing (FWM) process in a highly nonlinear fiber (HNLF) using linearly
chirped rectangular pump pulses. The pump pulses are obtained from a 10 GHz MLL at 1557 nm followed by spectral broadening in a dispersion-flattened HNLF (DF-HNLF). To obtain linearly chirped pumps, the output spectrum of the DF-HNLF is filtered using a wavelength selective switch (WSS), and each pump is subsequently propagated in an appropriate length of dispersive fiber. Pump1 is dispersed in 1700 m of SMF to achieve a chirp rate \( K = 0.055 \text{ ps}^2 \) for conversion of the 11.5 ps temporal spacing to a 100 GHz frequency grid (\( K < \frac{2 \pi A \omega^2}{\Delta \nu} \) due to the insertion of a GI). As the data-signal is phase-conjugated after the first FWM process, pump2 is dispersed in 290 m of dispersion compensating fiber (DCF) having the opposite dispersion value of the 1700 m SMF, in order to achieve the same chirp rate \( K \). The central wavelengths of the pumps are set at 1562 nm. The first FWM output is shown in Fig. 3 (c). After extraction with a 14 nm OBPF, the idler is dispersed in 115 m DCF, then combined with pump2 and coupled into HNLF2 for the second FWM process. The resulting spectrum is also shown in Fig. 3 (c). The generated idler is the Nyquist-WDM super-channel converted from the OFDM signal. Fig. 3 (d) shows a zoom-in on the idler, where 8 rectangular Nyquist-WDM channels with 100 GHz spacing can be observed, and the SE remains at 0.8 symbol/s/Hz after conversion.

In the receiver, a 100 GHz rectangular OBPF is used to demultiplex each Nyquist-WDM channel as shown in Fig. 3 (d). Each 80-Gbit/s Nyquist-WDM channel contains eight 10-Gbit/s Nyquist tributaries. Fig. 4 (a) shows the waveform of a WDM-demultiplexed 80-Gbit/s Nyquist-WDM channel, where 8 minimum- ISI positions with 11.5-ps spacing are clearly observed. These minimum- ISI positions correspond to the nulls of the sinc-shaped tributary waveforms. The minimum- ISI position in each 11.5-ps tributary time-slot is then sampled in the NOLM, using a 1.3-ps wide control pulse for gating. Finally, the BER of each tributary is measured in a 10-Gbit/s pre-amplified DPSK receiver including a delay line interferometer (DLI) and a balanced photo-detector.

Fig. 4 (b) shows the receiver sensitivities at BER=10^-9, which is achieved for all 64 10-Gbit/s DPSK tributaries within the 8 converted Nyquist-WDM channels. As a BER curve example, one of the best Nyquist tributaries (W5-T5: tributary 5 from Nyquist-WDM channel 5) is plotted in Fig. 4 (c). BER<10^-9 is achieved with a receiver sensitivity of -35.7 dBm. For reference, the BER curves of the same tributary without adjacent tributaries and neighboring Nyquist-WDM channels are also shown in Fig. 4 (c) respectively, as well as a 10 Gbit/s DPSK baseline. The power penalty for the best tributary is 5 dB for the full system (including all-optical OFDM generation, OFDM to Nyquist-WDM conversion and optical sampling), of which 2.3 dB is due to the ISI from the adjacent tributaries and 2 dB is due to the neighboring Nyquist-WDM channels.

4. Conclusion

We have proposed a novel all-optical ultra-high-speed OFDM to Nyquist-WDM conversion scheme based on complete OFT. An 8-subcarrier 640 Gbit/s DPSK OFDM signal was simultaneously converted to eight 80-Gbit/s Nyquist-WDM channels, achieving BER<10^-9 performance. The SE remains unchanged at 0.8 symbol/s/Hz after conversion, which is close to the binary single channel theoretical limit of 1 symbol/s/Hz. In principle, this OFT-based technique can also be employed for Nyquist-WDM to OFDM conversion, and it could therefore provide a simple and energy-efficient link between spectrally efficient OFDM and Nyquist-WDM systems.

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References